

## Performance Investigation of a Permanent Magnet DC Machines using Robust Control Technique

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**Key words:** Permanent magnet DC motor, Permanent magnet DC generator, H<sub>2</sub> optimal controller, H<sub>∞</sub> synthesis controllers

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**Abstract:** In this study, the design and performance investigation of the permanent magnet machines have been done to increase the mechanical and electrical outputs improvement of the systems. A Permanent Magnet DC Motor (PMDM) and Generator (PMDG) have been modelled and designed to improve the angular position and generated current respectively. In this research, augmentation based H<sub>2</sub> optimal and H<sub>∞</sub> synthesis controllers have been designed as a controller for the two systems and a comparison between the proposed controller for tracking a reference inputs and a promising results have been obtained.

## INTRODUCTION

The permanent magnet machines have the benefit of getting an excitation from the permanent magnet which is found at the stator part of the machines in order to get some efficiency benefit for that. Induction machines have a good and regulated fluxes that's why it is helpful to optimize the efficiency. Both systems are used for variable-speed drive. The specification of the performance and efficiency of the permanent magnet machine better has better cost minimization function with optimal for the range and performance target. In this study, the modeling design and control of a permanent magnet DC motor and generator for improving the angular speed and generating current have been done using robust control theory<sup>[1,2]</sup>

## MATERIALS AND METHODS

### Mathematical modelling

**Permanent magnet DC motor modelling:** Consider the cross section of a motor shown in Fig. 1 that has several loops of wire<sup>[3]</sup>.

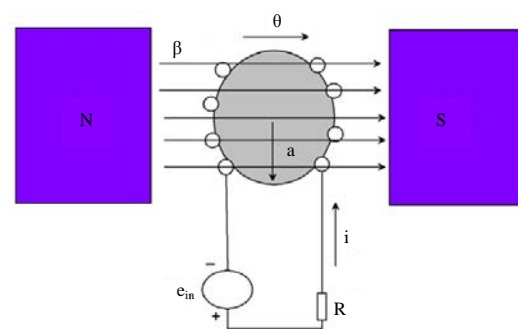


Fig. 1: PMDM cross section

The static magnetic field formed by an electromagnet with strength  $\beta$  going from left to right (North to South). The thickness of the magnet is defined by  $\ell$ . The loops of the wire around a "rotor" which is an iron cylinder with radius "a" that is free to rotate about its center. There are "n" rounds of wire with current going into the page on the right side and comes out of the page to the right<sup>[4]</sup>.

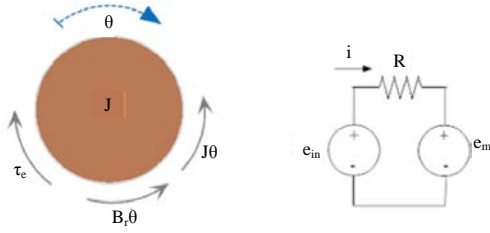


Fig. 2: Mechanical and electrical systems

The rotational angle,  $\theta$  is positive in the clockwise direction, the resistance of the wire is given by  $R$ . The moment of inertia of the rotor, “ $J$ ” with friction of rotation “ $B_r$ ”. The mathematical model for this system is derived by separating the system as mechanical system and the electrical system as shown in Fig. 2. The torque developed in the loop of wire is given by:

$$\begin{aligned} \tau_c &= 2nail\beta = \alpha i \\ \alpha &= 2nal\beta \end{aligned} \quad (1)$$

With a back emf:

$$e_m = 2n\omega a\beta = \alpha\omega \quad (2)$$

The electrical source is the input to the and the mechanical torque is the output of the motor. From the mechanical body diagram we get:

$$J\ddot{\theta} + B_r\dot{\theta} = \tau_c = \alpha i \quad (3)$$

And from the electrical free diagram we get:

$$e_{in} - iR = e_m = \alpha\omega \quad (4)$$

Substituting Eq. 3 for current into Eq. 4 yields:

$$e_{in} - \frac{J\ddot{\theta} + B_r\dot{\theta}}{\alpha} R = \alpha\omega = \alpha\dot{\theta} \quad (5)$$

Taking the Laplace transform with initial conditions set to zero and solve for the ratio of output to input yields:

$$\frac{\Theta(s)}{E_{in}(s)} = \frac{\alpha}{JR s^2 + (B_r R + \alpha^2)s}$$

The parameters of the PMDM is shown in Table 1. The transfer function numerically becomes:

$$\frac{\Theta(s)}{E_{in}(s)} = \frac{1}{4.44s^2 + 8.47s}$$

**Permanent magnet DC generator modelling:** A permanent magnet DC generator is typically the same

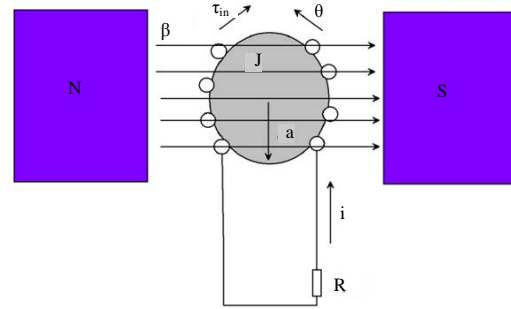


Fig. 3: PMDG cross section

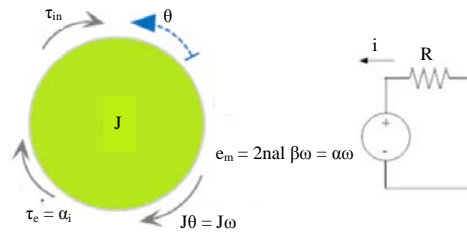


Fig. 4: Mechanical and electrical systems

Table 1: PMDM parameters

Parameters	Symbols	Values
Resistance of the wire	$R$	10 ohm
Moment of inertia of the rotor	$J$	0.8 kgm <sup>2</sup> /s <sup>2</sup>
Friction of rotation	$B_r$	1.2 Nms
Motor specification	$\alpha$	1.8

machine as a motor. The motor uses a mechanical input and generates an electrical output (Fig. 3). The free body diagram of the mechanical and electrical system is shown in Fig. 4. In this system, consideration of the direction of the induced torque is essential. The positive current is coming out of the page on the left side of the rotor and the field is to the right<sup>[5]</sup>. The back emf is in the positive direction with counterclockwise direction of the wires on the left side have positive velocity downward:

$$\begin{aligned} \tau_c + \tau_{in} + J\dot{\omega} &= 0 \\ \alpha i + \tau_{in} + J\dot{\omega} &= 0 \end{aligned} \quad (6)$$

$$\begin{aligned} e_m + iR &= 0 \\ \alpha\omega + iR &= 0 \end{aligned} \quad (7)$$

Taking the Laplace Transform of Eq. 6 and 7 and eliminating the angular velocity,  $\omega(s)$  yields:

$$\frac{I(s)}{T_{in}(s)} = \frac{\alpha}{JR s + \alpha^2}$$

The parameters of the PMDG is shown in Table 2. The transfer function numerically becomes:

Table 2: PMDG parameters

Parameters	Symbols	Values
Resistance of the Wire	R	15 ohm
Moment of inertia of the rotor	J	0.6 kgm <sup>2</sup> /s <sup>2</sup>
Motor specification	$\alpha$	2.9

$$\frac{I(s)}{T_{in}(s)} = \frac{1}{3.1s + 2.9}$$

**Proposed controllers design**

**Augmentations of the model with weighting functions:**

The weighted control structure of the systems is shown in Fig. 4 where W1(s), W2(s) and W3(s) are weighting functions. The assumption that G(s), W1(s) and W3(s) G(s) are all proper systems. The weighting function W3(s) is not required to be proper system. In the state space structure of the systems, the output vector  $y1 = [y1a, y1b, y1c]^T$  cannot be used directly to the control signal u. Clearly, Fig. 5 represents a more general picture of optimal and robust control systems. The design of the H2 optimal and H $\infty$  synthesis controllers is done by using the idea of the augmented state space model<sup>[6]</sup>.

**For the motor:** The weighting function W1m(s), W2m(s) and W3m(s) are chosen as:

$$W_{1m}(s) = \frac{s+1}{2s+5} \quad W_{2m}(s) = \frac{s+8}{3s+24} \quad W_{3m}(s) = 7$$

The H2 optimal controller become:

$$G_{mH_2} = \frac{3.1679s^2 + 11.3466s + 41.7504}{s^3 + 47.1659s^2 + 125.7358s + 28.2287}$$

The H $\infty$  synthesis controller become:

$$G_{mH_\infty} = \frac{3.26s^2 + 12.2s + 28.38}{s^3 + 47.2s^2 + 128.6s + 27.6}$$

**For the generator:** The weighting function W1g(s), W2g(s) and W3g(s) are chosen as:

$$W_{1g}(s) = \frac{s+4}{s+8} \quad W_{2g}(s) = \frac{s+12}{2s+14} \quad W_{3g}(s) = 9$$

The H2 optimal controller become:

$$G_{gH_2} = \frac{2.5736s + 12.0904}{s^2 + 5.2298s + 18.7438}$$

The H $\infty$  synthesis controller become:

$$G_{gH_\infty} = \frac{3.2096s + 18.3848}{s^2 + 8.6176s + 17.4376}$$

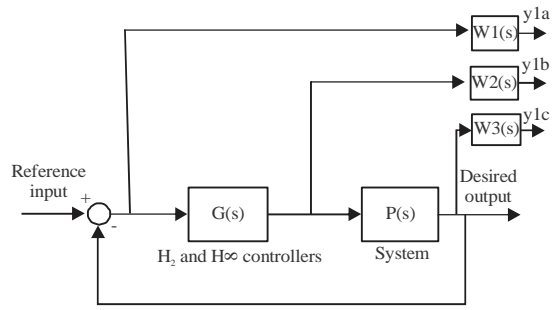


Fig. 5: Weighted control structure with the proposed controllers

**RESULTS AND DISCUSSION**

**Comparison of a PMDM with H2 optimal and H $\infty$  synthesis controllers for improving the angular position using a step reference angular position input:**

The angular position output of a PMDM performance analysis is done by simulating the system with the proposed controllers for a step input reference input and the simulation result is shown in Fig. 6.

The simulation result shows that the PMDM with H2 optimal controller improves the angular position in minimizing the percentage overshoot and the settling time better than the system with H $\infty$  synthesis controller<sup>[7]</sup>.

**Comparison of a PMDM with H2 optimal and H $\infty$  synthesis controllers for improving the angular position using a random reference angular position input:**

The angular position output of a PMDM performance analysis is done by simulating the system with the proposed controllers for a random input reference input and the simulation result is shown in Fig. 7.

The simulation result shows that the PMDM with H2 optimal controller improves the angular position in tracking the reference input signal with better amplitude than the system with H $\infty$  synthesis controller.

**Comparison of a PMDG with H2 optimal and H $\infty$  synthesis controllers for improving the generating current using a step reference current input:**

The angular position output of a PMDG performance analysis is done by simulating the system with the proposed controllers for a step input reference input and the simulation result is shown in Fig. 8.

The simulation result shows that the PMDG with H2 optimal controller improves the generating current in minimizing the percentage overshoot better than the system with H $\infty$  synthesis controller.

**Comparison of a PMDM with H2 optimal and H $\infty$  synthesis controllers for improving the generating current using a random reference current input:**

The angular position output of a PMDM performance analysis

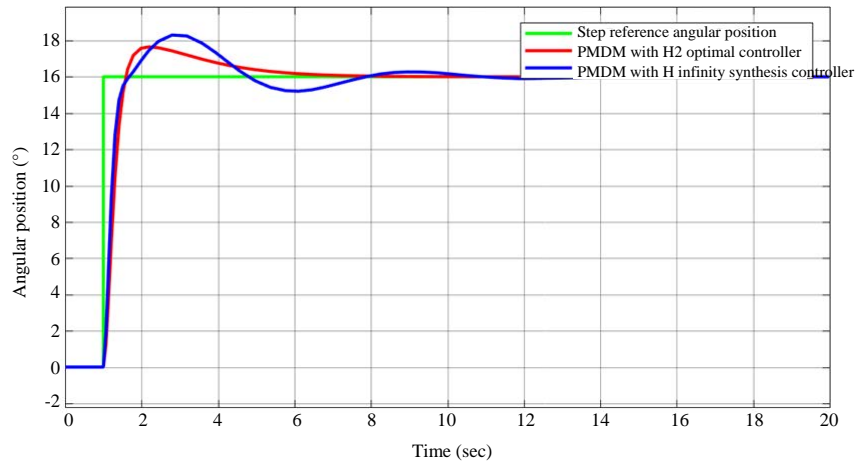


Fig. 6: Step response of the PMDM

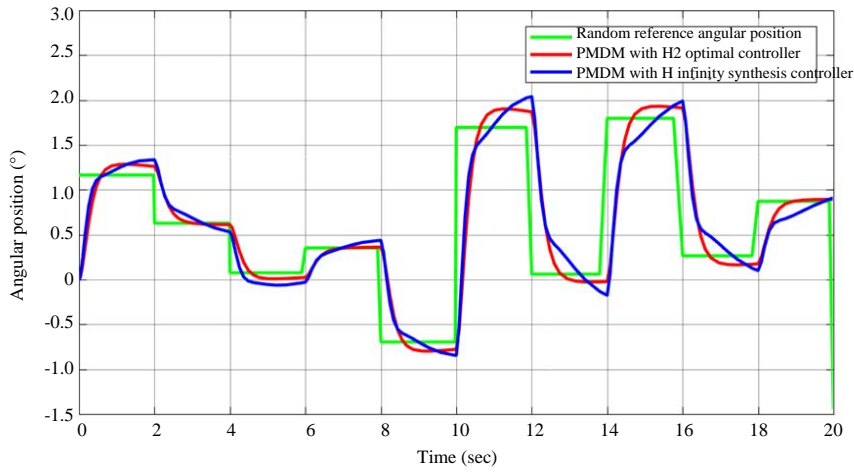


Fig. 7: Random response of the PMDM

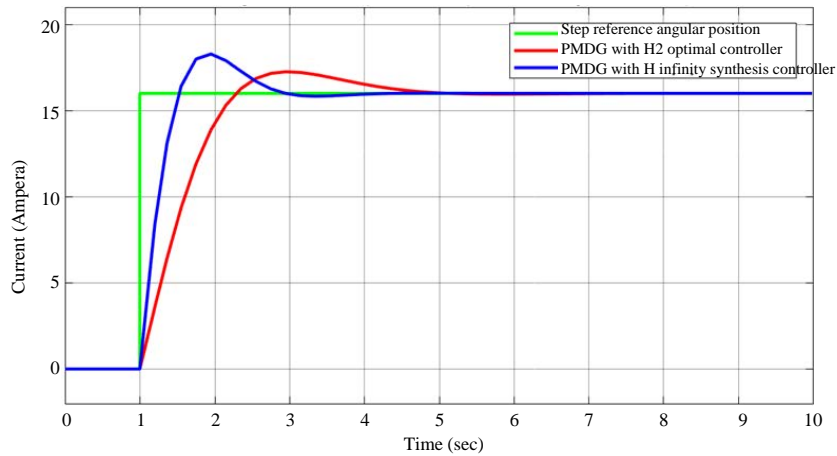


Fig. 8: Step response of the PMDG

is done by simulating the system with the proposed controllers for a random input reference input and the

simulation result is shown in Fig. 9. The simulation result shows that the PMDG with H2 optimal controller

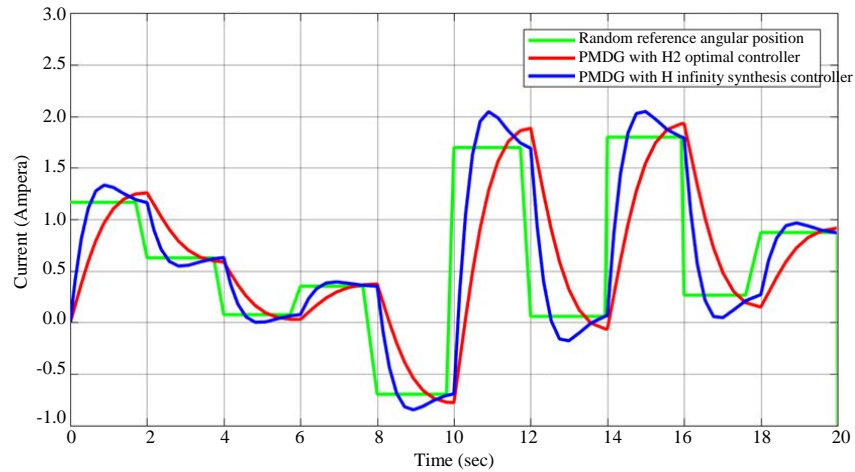


Fig. 9: Random response of the PMDG

improves the angular position in tracking the reference input signal with better amplitude than the system with  $H^\infty$  synthesis controller.

**CONCLUSION**

The modelling, design and control of a permanent magnet Dc motor for an improvement of angular position and a permanent magnet Dc generator for improving of the generating current using augmentation based  $H_2$  optimal and  $H^\infty$  synthesis controllers. A comparison of the proposed systems with the proposed controllers for the analysis of the performance improvement of the angular position and generating current using a step and random reference input signals. The simulation result of the PMDM for a step reference angular position input suggested that the PMDM with  $H_2$  optimal controller improves the angular position in minimizing the percentage overshoot and the settling time while the simulation result of the PMDM for a random reference angular position suggested that the PMDM with  $H_2$  optimal controller improves the angular position in tracking the reference input signal with better amplitude. The simulation results of the PMDG suggested that the PMDG with  $H_2$  optimal controller improves the generating current in minimizing the percentage overshoot for a step input reference current and the PMDG with  $H_2$  optimal controller improves the angular position in tracking the reference input signal with better amplitude.

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